



# Workshop on low-power integrated circuits dedicated to 3G mobile phones

A reconfigurable 4-GHz VCO for 3G multimode transceivers

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- Application : multimode 3G transceivers
- Frame : IST project MuMoR
- Objective : design a VCO with reconfigurability features
- Targeted modes :
  - UMTS W-CDMA FDD and TDD, RX & TX
- VCO design features :
  - fully on-chip integrated.
  - includes all the auxilliary blocks => get real-world data.
  - immunity to substrate and supply noise : architecture choice.
  - wide frequency tuning range : switched capacitors.
  - low power : automatic amplitude control (AAC) loop.



# Specifications

- VCO runs at twice the frequency.
- RX can be optimized for low power.
- TX must be optimized for low phase noise.
- Overall frequency range : 3800-4340 MHz

RX MODES				
Parameter	Mode	Value		Unit
		Min.	Max.	
Freq.	TDD band1	3800	3840	MHz
	TDD band2	4040	4050	MHz
	FDD	4220	4340	MHz
Noise @ 5 MHz	All		-97	dBc/Hz
Noise @ 10 MHz	All		-107	dBc/Hz
Noise @ 15 MHz	All		-119	dBc/Hz
Pushing	All		5	MHz/V
Supply voltage	All	2.7	2.9	V
Temperature	All	-20	+85	°C

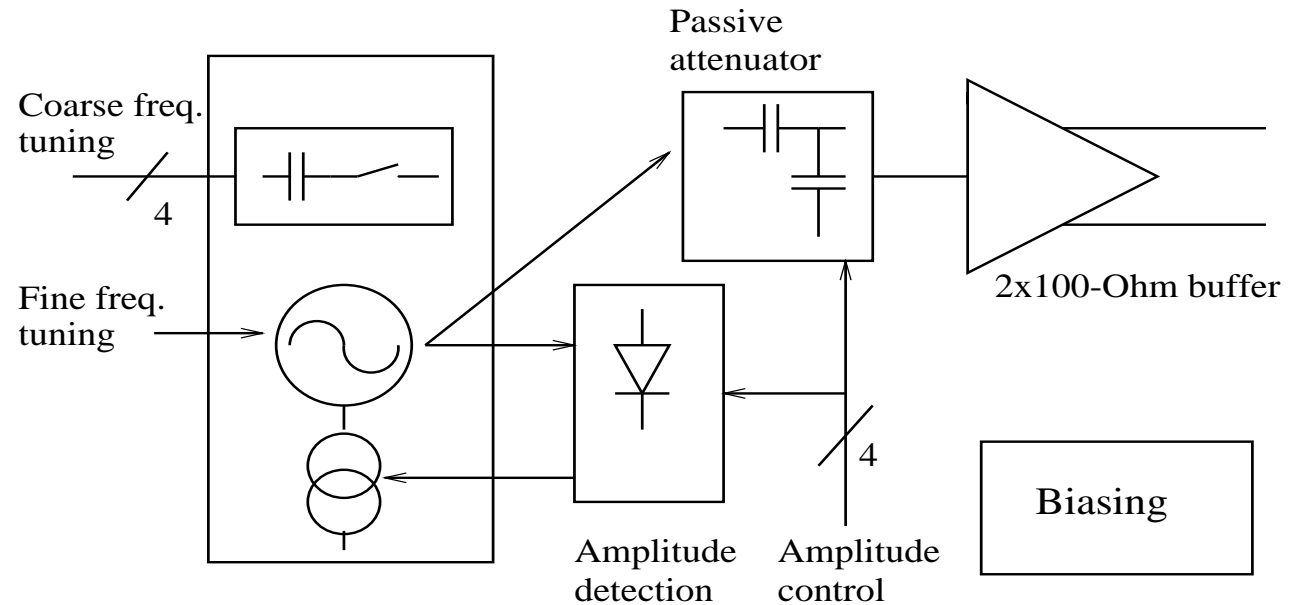
TX MODES				
Parameter	Mode	Value		Unit
		Min.	Max.	
Freq.	TDD band1	3800	3840	MHz
	TDD band2	4040	4050	MHz
	FDD	3840	3960	MHz
Noise @ 3 MHz	All		-120	dBc/Hz
Noise @ 12.5 MHz	All		-123	dBc/Hz
Noise @ 20 MHz	All		-145	dBc/Hz
Noise @ 85 MHz	All		-148	dBc/Hz
Pushing	All		5	MHz/V
Supply voltage	All	2.7	2.9	V
Temperature	All	-20	+85	°C



# Overall architecture

Three reconfigurable features :

- coarse freq. tuning to switch between modes.
- amplitude control to keep power as low as possible.
- output level maintained constant to ease integration in system-on-chip.

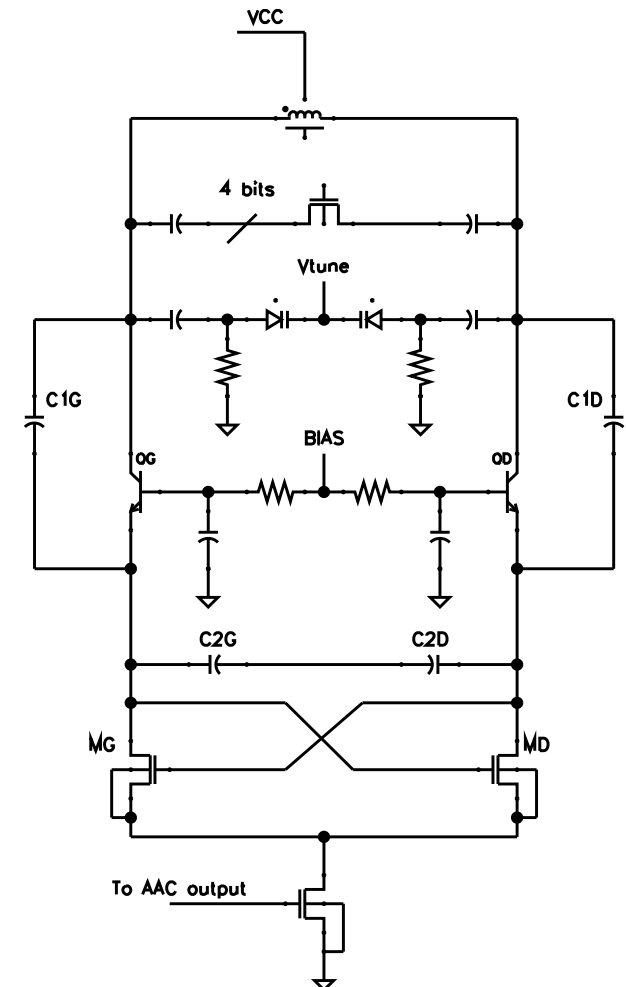




## Core oscillator

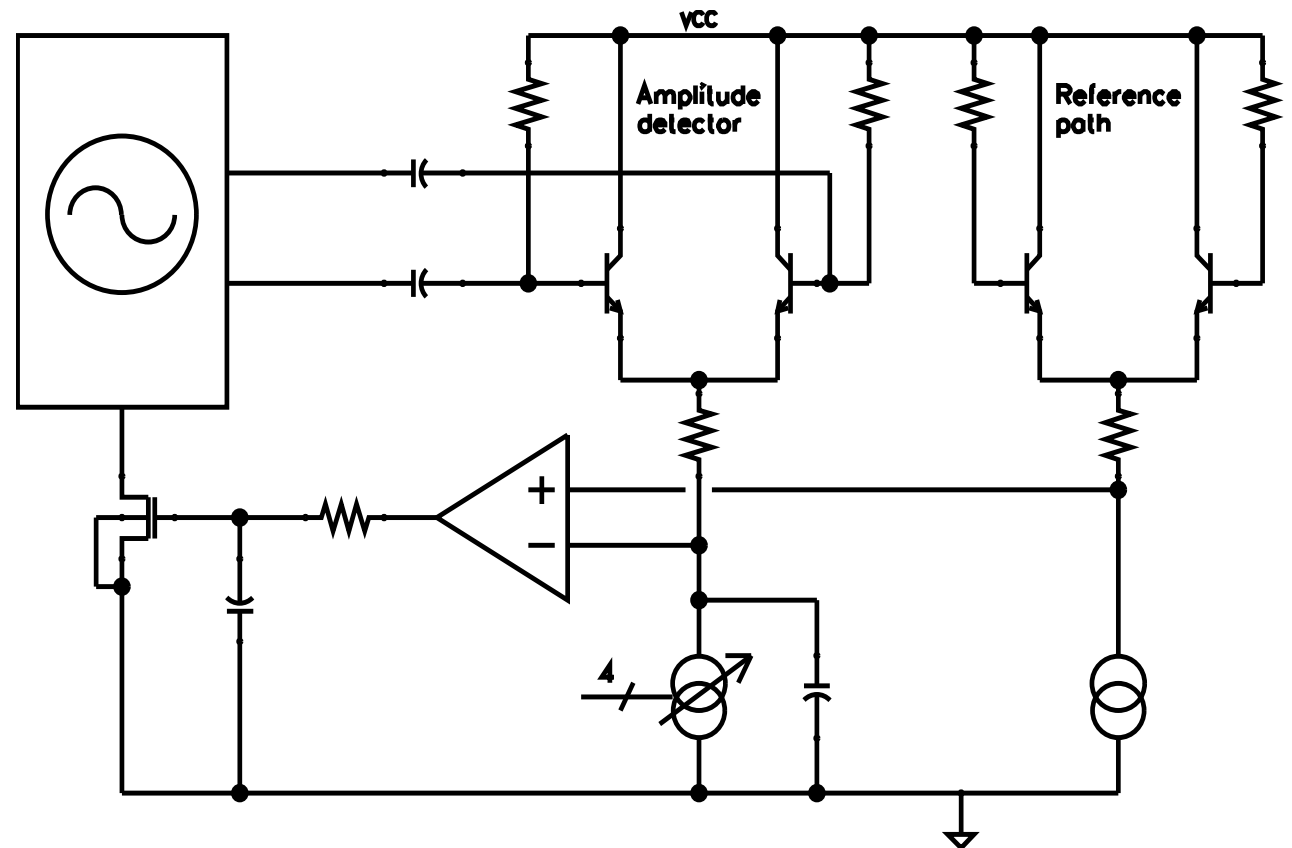
- recently published architecture [1] which combines low phase noise and low power :
  - each of the bipolars is connected in Colpitts configuration for best phase noise.
  - cross-coupled MOSFETs switch current from the source twice by period to reduce current consumption.
- Differential structure rejects common mode noise coming from substrate.
- Tail source provides isolation from supply.

[1] R. Aparicio, A. Hajimiri, "A CMOS differential noise-shifting Colpitts VCO", ISSCC 2002.





- VCO 4-GHz output is rectified and low-pass filtered.
- Resulting voltage is compared to reference.
- Good low-pass filtering is needed to get good low-offset phase noise (i.e. @ 1 MHz and less).
- settling time for a full-range step is 2-3  $\mu$ s.

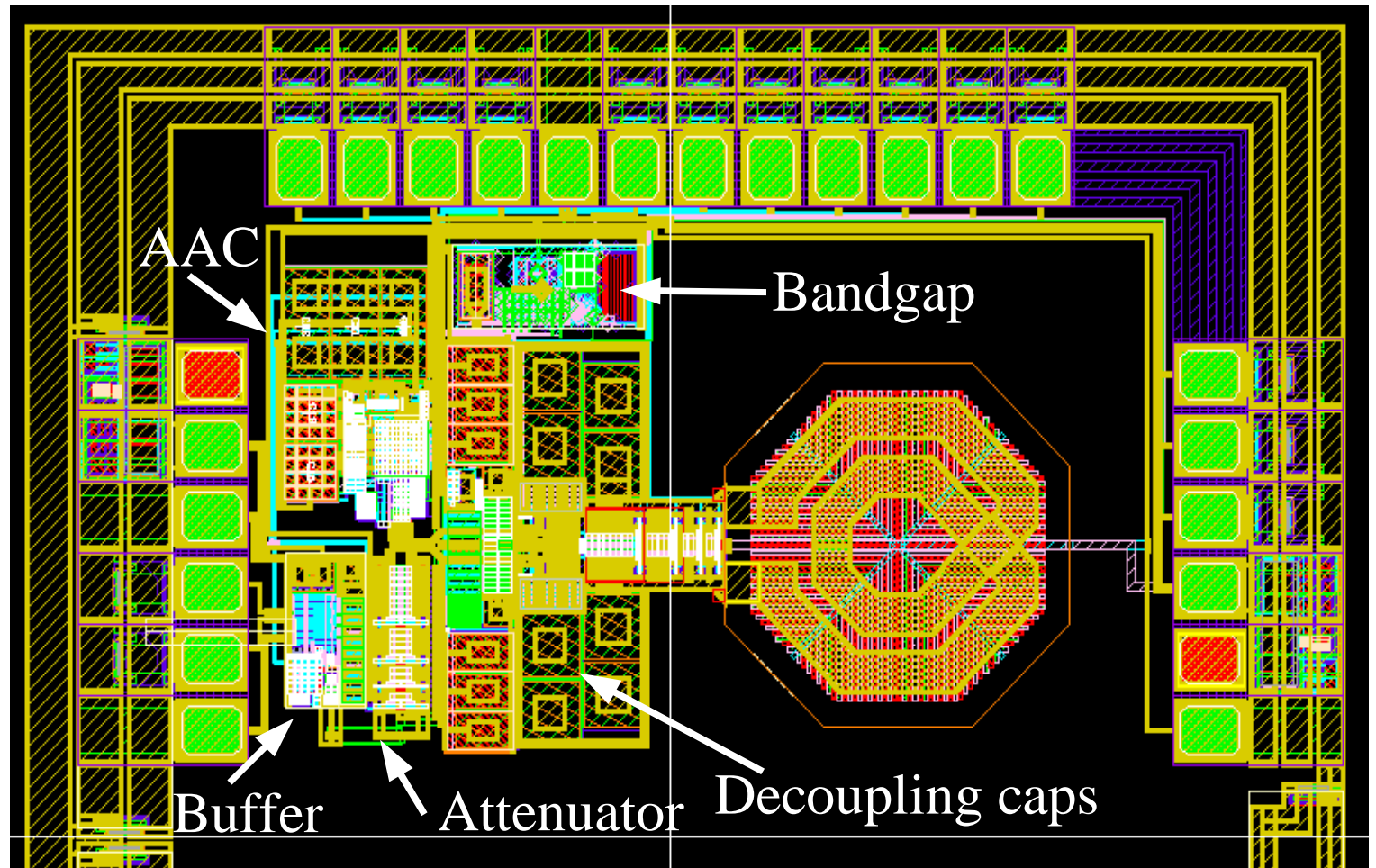




# Layout

Size without bonding pads :  
1000 x 600  $\mu\text{m}$ .

Output buffer has its own supply.

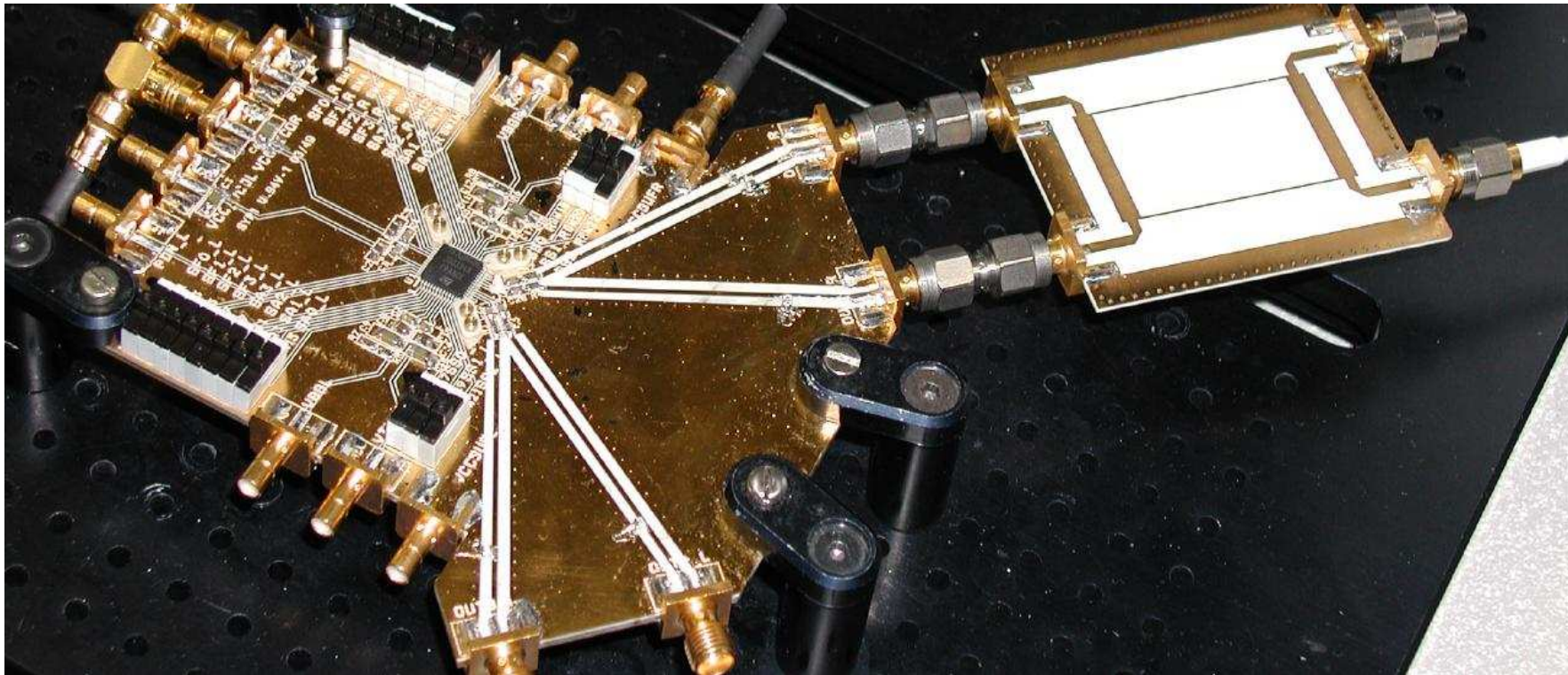




MUMOR

## Test setup

- Rogers 4350 board + hybrid coupler (diff -> single) on outputs.
- $2 \times 100 \Omega \rightarrow 2 \times 50 \Omega$  matching made on PCB.

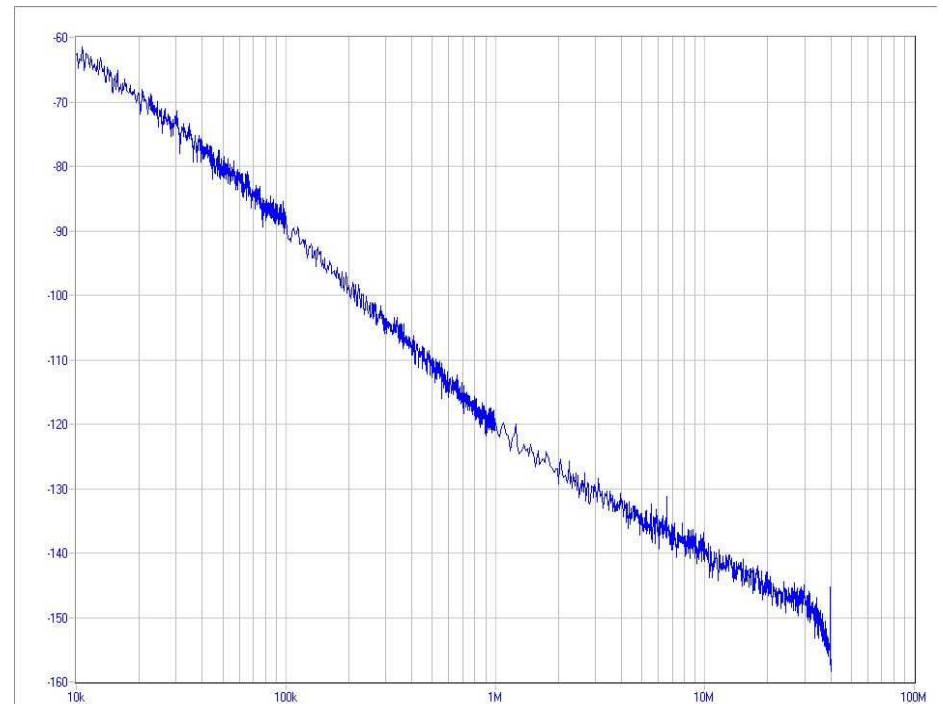
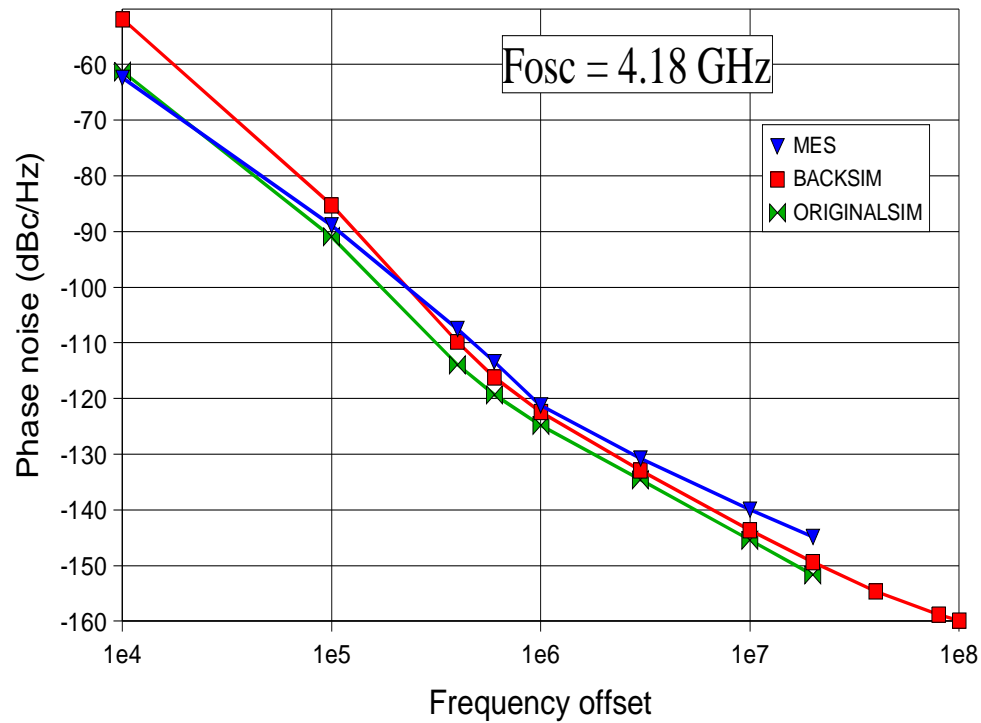




# Phase noise measurement

Phase noise measured on PN9000 testbench.

Frequency : 4.18 GHz  
Core VCO current : 8 mA





## Comparing simulated and measured phase noise

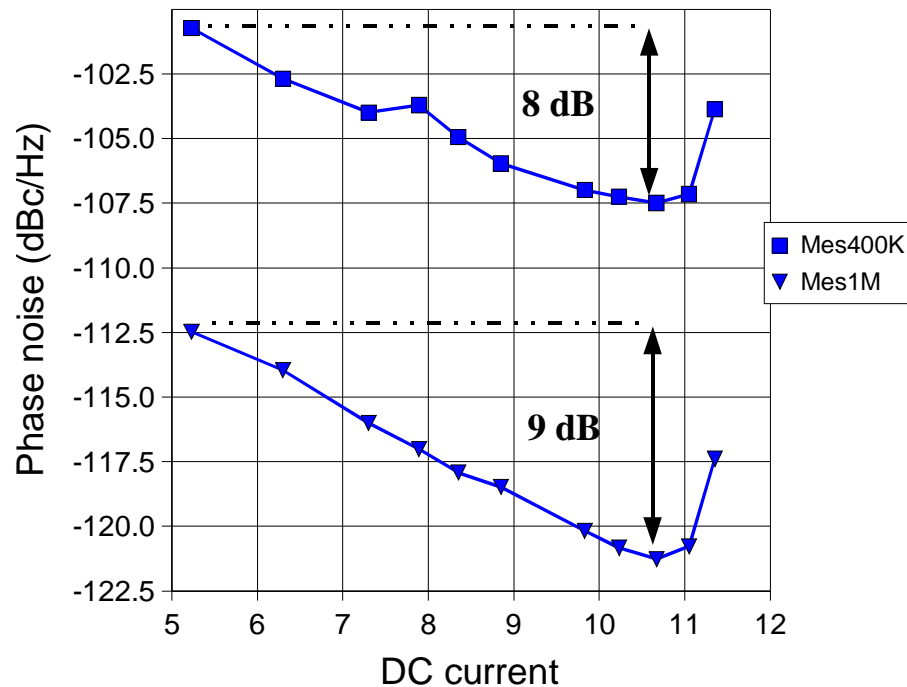
- Limitations :
  - AAC loop can not deliver currents high enough :
    - ✓ bandgap resistor was changed to allow more measurements.
    - ✓ but bandgap and biasing blocks are more noisy than they should.
  - bad substrate connections => some instability.
- Comparing back-simulations and measurements :
  - Phase noise difference is quite good (2-3 dB) at low offsets (< a few MHz).
  - Uncertainty on the substrate modelling which impact the simulation accuracy.
  - At higher offsets the bigger difference can be explained by the noise of the supply which improves significantly above 5 MHz and which is not modelled.
- Comparing original simulation and measurements :
  - phase noise behavior over all offset frequencies is well modelled.



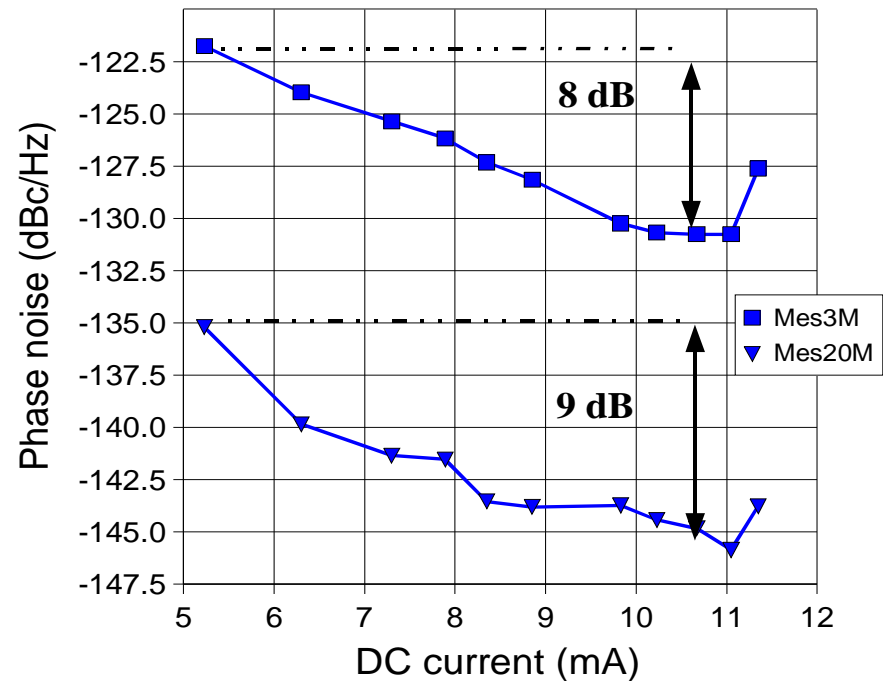
# Switching between low-power and low-noise configurations

- When current is halved high-offset phase noise is degraded by 8 to 9 dB.

### Low-offset phase noise



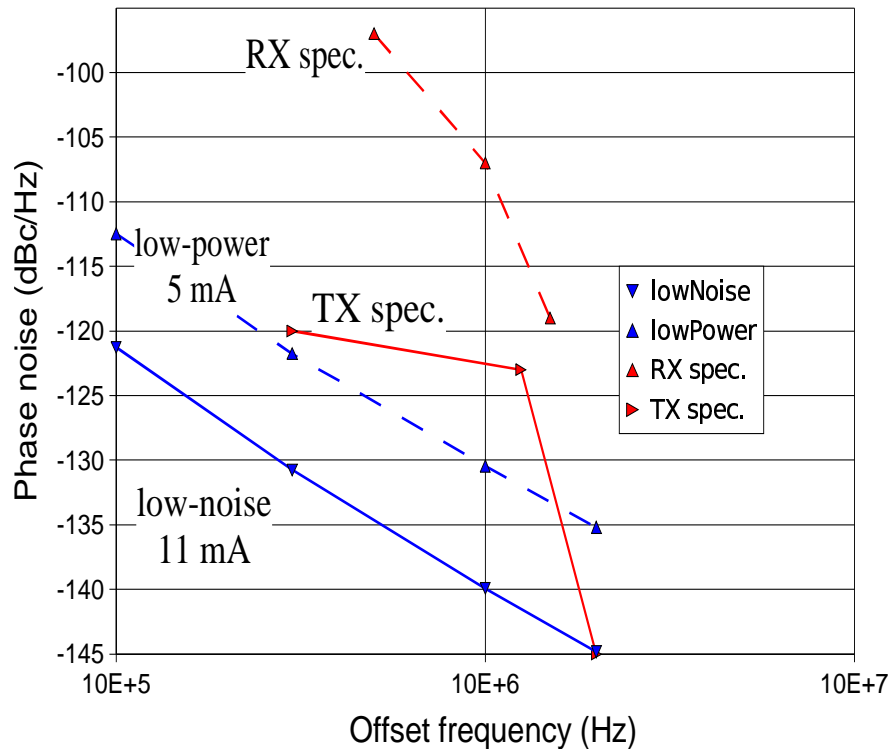
### High-offset phase noise



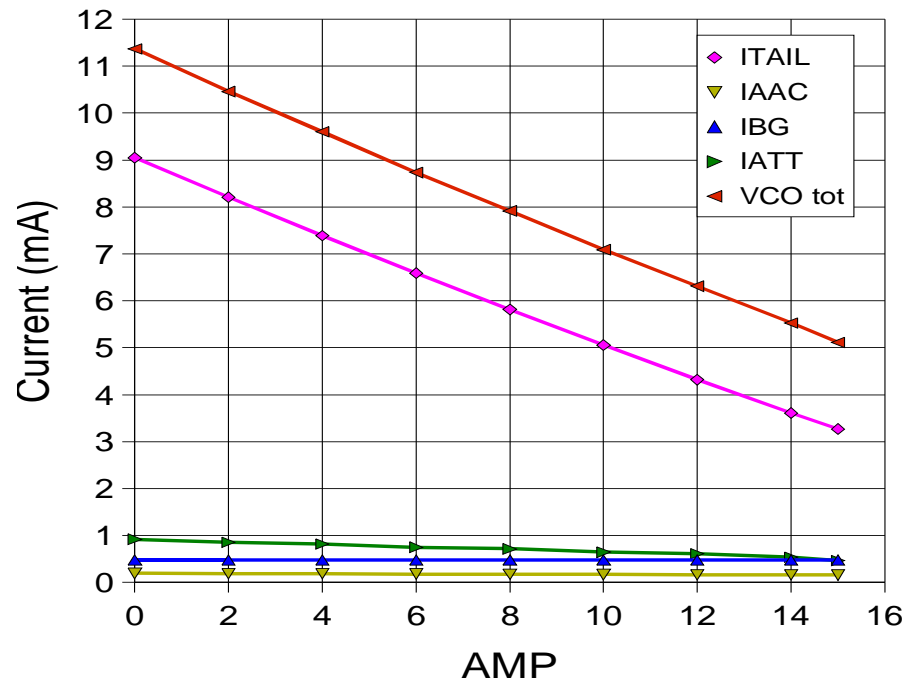


# Low-noise and low-power configurations

- Current consumption can be halved from TX to RX.
- Auxiliary blocks consumption is moderate compared to the VCO.



## Current sharing





# Phase noise figures

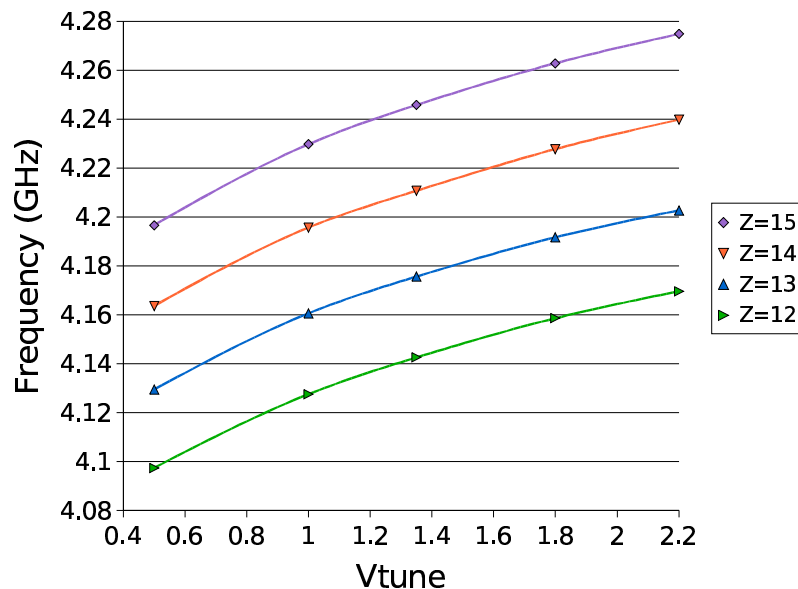
Freq. offset	Meas.	VCO		UMTS W-CDMA	
		Back-simul.	Original simul.	RX	TX
@ 10KHz	-62.4	-51.9	-61.3		
@ 100KHz	-88.9	-85.3	-90.9		
@ 400KHz	-107.5	-109.8	-113.9		
@ 600KHz	-113.4	-116.2	-119.3		
@ 1MHz	-121.3	-122.4	-124.8		
@ 3 MHz	-130.8	-132.9	-134.6		-120.0
@ 5 MHz	-135.2	-137.3	-139.0	-97.0	
@ 10MHz	-139.9	-143.6	-145.4	-107.0	
@ 12.5MHz	-141.8	-145.5	-147.3		-123.0
@ 15MHz	-143.4	-147.1	-148.9	-119.0	
@ 20MHz	-144.8	-149.4	-151.7		-145.0
@ 85MHz		-158.8			-148
@ 100MHz		-159.8			



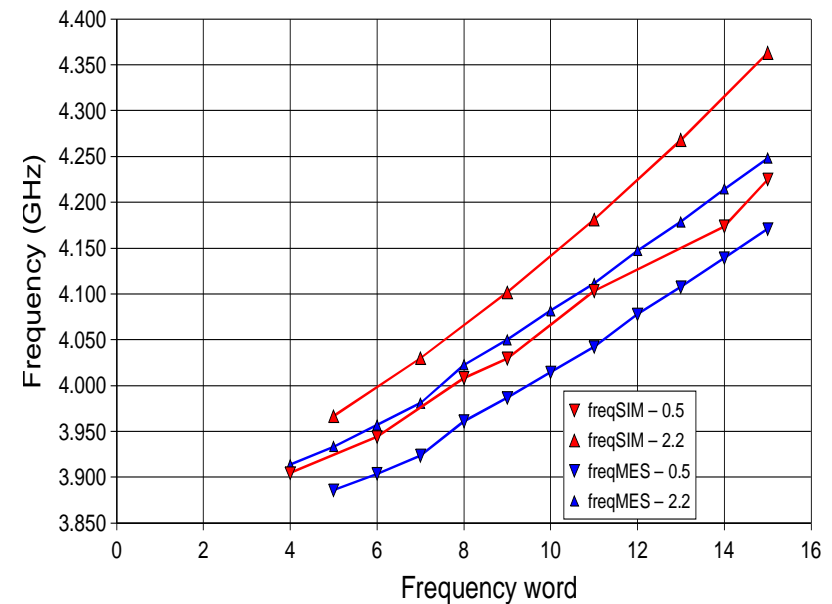
# Frequency tuning

- Full frequency range not measured :
  - the VCO needs more current at low frequency... but the AAC loop cannot provide it.
  - back-simulations reproduce these phenomenons.
- Frequency range can be widened with additional switched capacitors.

### Oscillation frequency vs Vtune

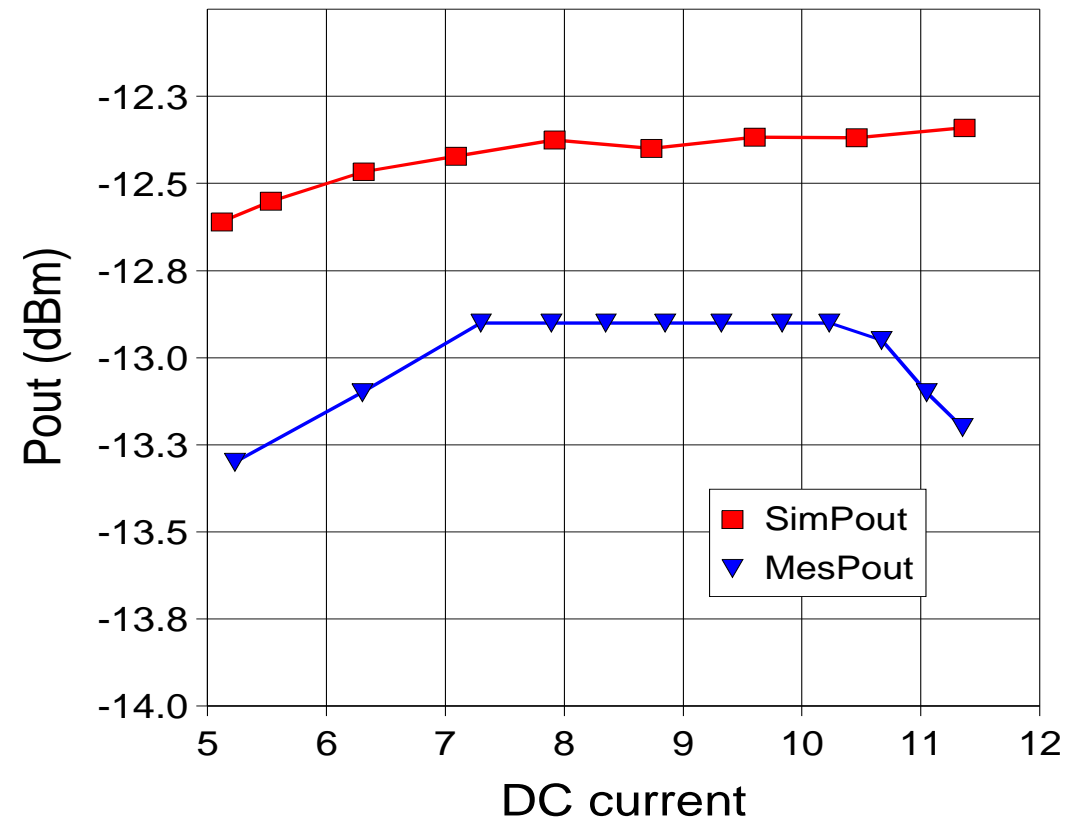


### Frequency coarse and fine tuning



- Output level remains within 0.5 dB when switching between low-power and low-noise configurations.
- The combination of the amplitude control and the variable attenuator performs well.

## Pout vs current

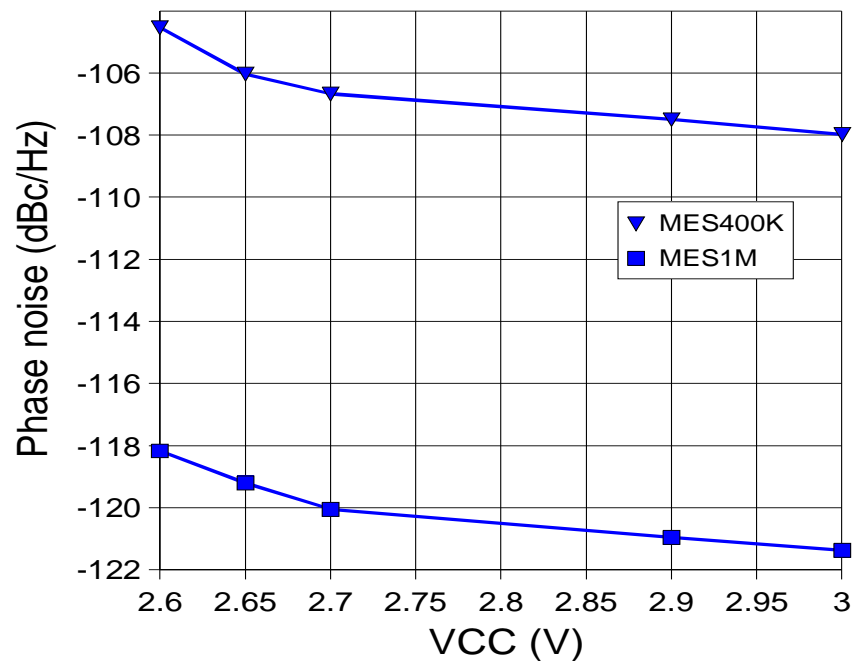




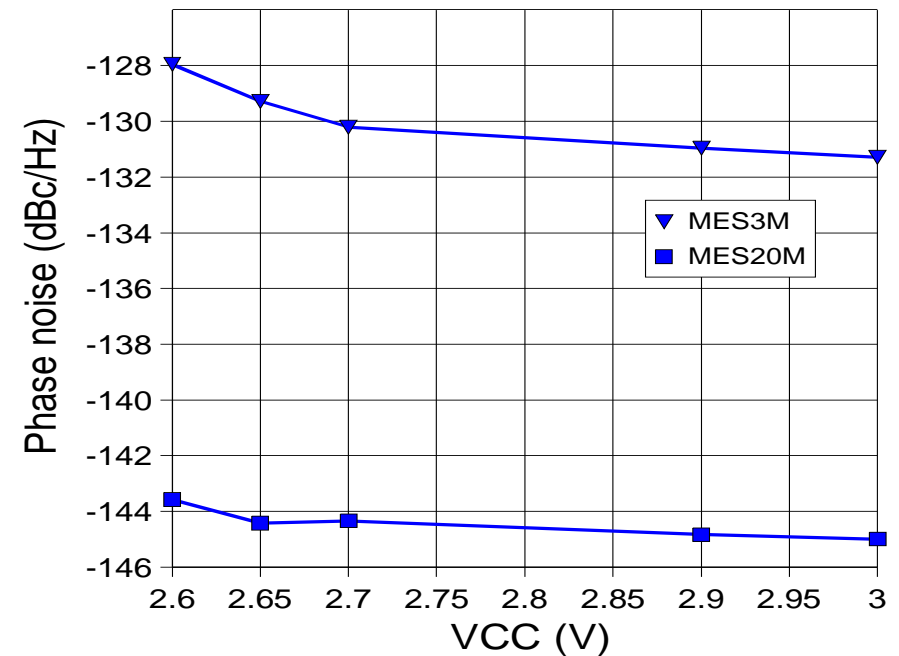
## Varying supply voltage

- Phase noise loosely depends on  $V_{CC}$ .

### Low-offset phase noise vs. VCC



### High-offset phase noise vs. VCC

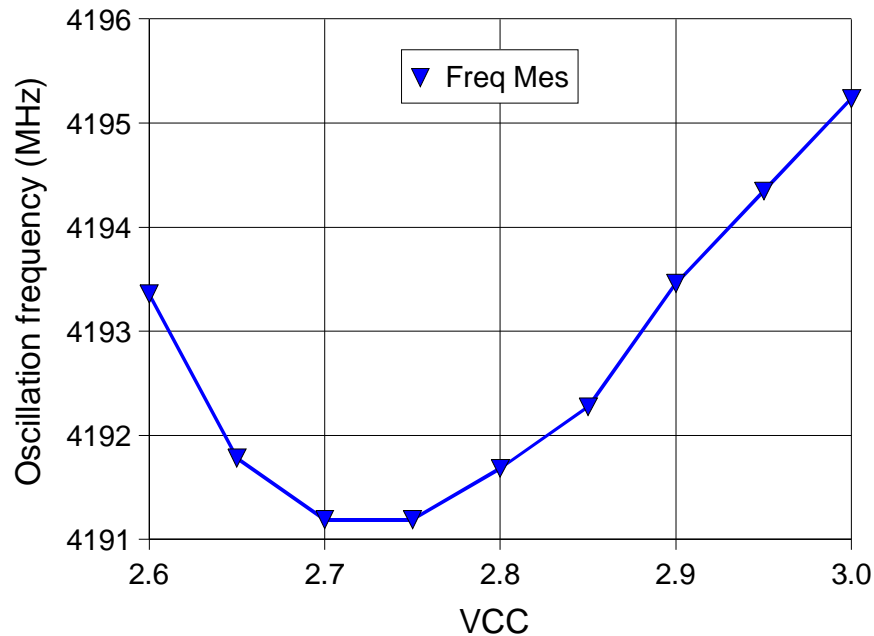




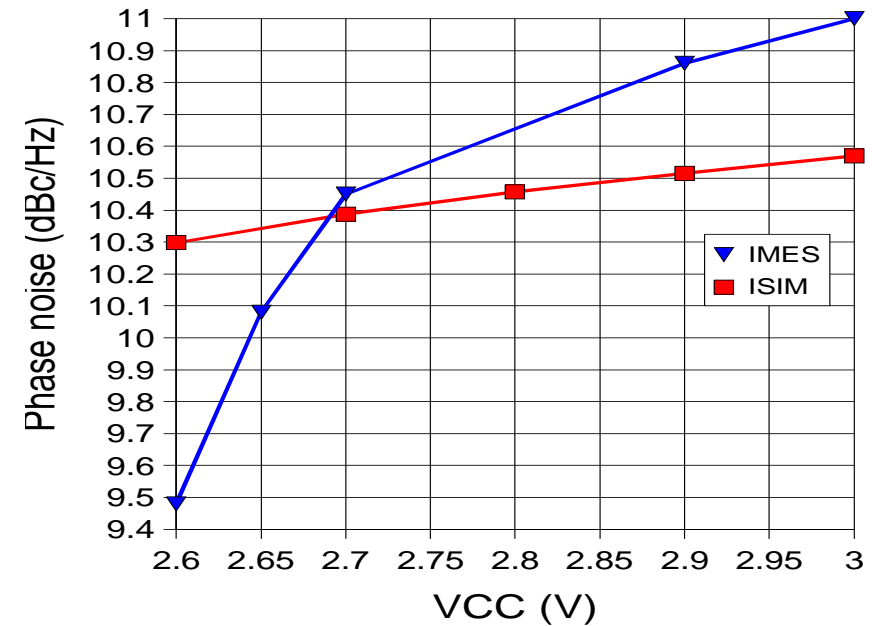
## Varying supply voltage

- Low frequency pushing
- Current consumption loosely depends on  $V_{CC}$

### Frequency pushing



### Total VCO current vs. VCC





- UMTS W-CDMA TX & RX noise specifications from MuMoR fulfilled.
- benefits of reconfiguration demonstrated :
  - low-power or low-noise modes possible with the same VCO.
- differential Colpitts architecture demonstrates good performance :
  - phase noise and current are comparable to classical differential pair.
  - FOM = -180 dBc/Hz (measured) and -185 dBc/Hz (simulated).